



## TPS4100A

### FOUR CHANNEL BRIDGED OUTPUT POWER STAGE

Technical Information

Revision 1.1 - November 2005

#### GENERAL DESCRIPTION

The TPS4100A is a 4 channel bridged (16 power transistors) output power stage. The TPS4100A accepts 5V CMOS logic signals from a Class-T processor, such as TCD6001, to create a high fidelity 4-channel audio amplifier. The TPS4100A has been designed specifically for automotive head unit applications operating on a single 10-26V supply.

#### APPLICATIONS

- Automotive Head Units and Trunk Amplifiers
- DVD Receivers
- Multimedia Speaker Systems

#### BENEFITS

- 4-channel output stage – with integrated driver and FETs - in a single 32-pin SSIP package
- Low external component count
- Single-supply operation

#### FEATURES

- Four H-Bridge outputs
- High Efficiency
- High Power @25.0V
  - 100W<sub>sat. sq. wave</sub> @ 4Ω
- High Efficiency
  - 88% @ 100W 4Ω
- AM “Low EMI” mode with connection to appropriate Class-T controller
- Mates seamlessly with TCD6001 Digital Input Class-T controller
- Mute and Stand-By function
- Protection Modes:
  - Output Short to VPP and Ground
  - Output Short across Load
  - Load Dump Protection
  - Over-/Under-Voltage Protection
  - Over-current Protection
  - Over-temperature Protection
  - Fortuitous Open Ground



**Absolute Maximum Ratings** (Note 1)

SYMBOL	PARAMETER	Value	UNITS
V <sub>PP</sub>	Supply Voltage (V <sub>PP</sub> )	33	V
V <sub>PP</sub> <sub>MAX</sub>	Peak Supply Voltage (t <sub>≤</sub> 50ms)	60	V
V <sub>PP</sub> <sub>AM</sub>	Supply Voltage in AM Mode (Note 2)	18	V
V <sub>IN</sub> <sub>RANGE</sub>	Voltage Range for Input Section Pins (Note 3) Inputs (Pins 1-11)	-0.5 to 5.5	V
T <sub>STORE</sub>	Storage Temperature Range	-55 to +150	°C
I <sub>R</sub>	Repetitive Peak Output Current	14	A
T <sub>J</sub>	Maximum Junction Temperature	150	°C
P <sub>D</sub>	Total Power Dissipation (T <sub>case</sub> = 70°C)	80	W
ESD	ESD Susceptibility - Human Body Model (Note 4)	2k	V
ESD	ESD Susceptibility – Machine Model (Note 5)	200	V

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.

See the table below for Operating Conditions.

Note 2: Supply voltage is limited in AM Mode due to additional power dissipation of output stage when operating in Class B mode. Please note that the TPS4100A is still fully protected from load dump transients in AM Mode.

Note 3: The input section pins (pins 1-9, 11,12) should not be connected to voltages over 5.5V with respect to pin 10 (AGND). Please note that pins 5, 11, and 12 are outputs and can be damaged if a voltage is forced externally.

Note 4: Human body model, 100pF discharged through a 1.5KΩ resistor.

Note 5: Machine model, 220pF – 240pF discharged through all pins.

**Operating Conditions** (Note 6)

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNITS
V <sub>PP</sub>	Supply Voltage (Note 6)	10	14.4	26	V
V <sub>PP</sub> <sub>AM</sub>	Supply Voltage for AM Mode (Note 7)	10	14.4	16	V
T <sub>A</sub>	Operating Free Air Temperature Range	-40	25	85	°C

Note 6: Recommended Operating Conditions indicate conditions for which the device is functional.

See Electrical Characteristics for guaranteed specific performance limits.

Note 7: Supply voltage is limited in AM Mode due to additional power dissipation of output stage when operating in Class B mode, as opposed to Switching Mode.

**Thermal Characteristics**

SYMBOL	PARAMETER	Value	UNITS
θ <sub>JC</sub>	Junction-to-case Thermal Resistance	1.0	°C/W
θ <sub>JA</sub>	Junction-to-ambient Thermal Resistance (still air)	20	°C/W

**Electrical Characteristics** (Note 8)

$T_A = 25\text{ }^\circ\text{C}$ . Unless otherwise noted, the supply voltage is  $V_{PP}=14.4\text{V}$ . See Application/Test Circuit.

SYMBOL	PARAMETER	Conditions	MIN.	TYP.	MAX.	UNITS
$I_{STBY}$	Stand-By Current	$V_{SLEEPB} < 0.15\text{V}$		100	200	$\mu\text{A}$
$V_{IL}$	Stand-By On Threshold Voltage	SLEEPB Low (amp off)			0.5	V
$V_{IH}$	Stand-By Off Threshold Voltage	SLEEPB High (amp on)	2.3			V
$V_{IL}$	Mute-On Threshold Voltage	MUTEB Low			1	V
$V_{IH}$	Mute-Off Threshold Voltage	MUTEB High	2.7			V
$V_{IL}$	Yn/YnB Low Threshold Voltage				0.6	V
$V_{IH}$	Yn/YnB High Threshold Voltage		2.7			V
$V_{OH}$	Fault Reporting Logic Output High Voltage	Open Drain Output	3.5			V
$V_{OL}$	Fault Reporting Logic Output Low Voltage	$R_{FAULT} = 51\text{K}\Omega$			1	V
$V_{IH}$	AM Mode On Threshold Voltage	AM pin High	2.7			V
$V_{IL}$	AM Mode Off Threshold Voltage	AM pin Low			1	V
$I_{AM}$	AM Mode Pin Input Current				1	$\mu\text{A}$

Note 8: Minimum and maximum limits are guaranteed but may not be 100% tested.

**Performance Characteristics** (Note 8)

$T_A = 25\text{ }^\circ\text{C}$ . Unless otherwise noted,  $R_L = 4\Omega$ . Measurement Bandwidth = 20kHz. All specifications shown are applicable only when the TPS4100A is used in conjunction with the TCD6001 Class-T Controller. See Application/Test Circuit of TCD6001 data sheet for additional information.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNITS
$P_{OUT}$	Output Power (Continuous power/ channel)	$V_{PP}=25\text{V}$ saturated sq. wave	95	112		W
		$V_{PP}=25\text{V}$ THD+N=10%		80		W
		$V_{PP}=25\text{V}$ THD+N=1%		65		W
		$V_{PP}=20\text{V}$ saturated sq. wave		74		W
		$V_{PP}=20\text{V}$ THD+N=10%		51		W
		$V_{PP}=20\text{V}$ THD+N=1%		40		W
		$V_{PP}=14.4\text{V}$ sat. sq. wave, $R_L = 2\Omega$		63		W
		$V_{PP}=14.4\text{V}$ THD+N=10%, $R_L = 2\Omega$		42		W
		$V_{PP}=14.4\text{V}$ THD+N=1%, $R_L = 2\Omega$		33		W
		$V_{PP}=14.4\text{V}$ saturated sq. wave		39		W
		$V_{PP}=14.4\text{V}$ THD+N=10%		26		W
		$V_{PP}=14.4\text{V}$ THD+N=1%		21		W
$\eta$	Power Efficiency	$V_{PP}=25\text{V}$ , 4 x 100W sat sq wave		88		%

**AM Mode** (Notes 8, 9)

$T_A = 25\text{ }^\circ\text{C}$ . Unless otherwise noted, the supply voltage is  $V_{PP}=14.4\text{V}$ ,  $R_L = 4\Omega$ . Measurement Bandwidth = 20kHz. See Application/Test Circuit.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNITS
$I_{OCD}$	Over-current detect		5.5			A
$P_{out}$	Output Power (Note 7)	$V_{PP}=16\text{V}$ , THD+N=10%		20		W
		$V_{PP}=14.4\text{V}$ , THD+N=10%		16		W

Note 9: The TPS4100A heat sinking in AM Mode must be increased (as compared to Class-T mode) to sustain the typical output numbers. This is due to the lower efficiency of Class B output stage operation.

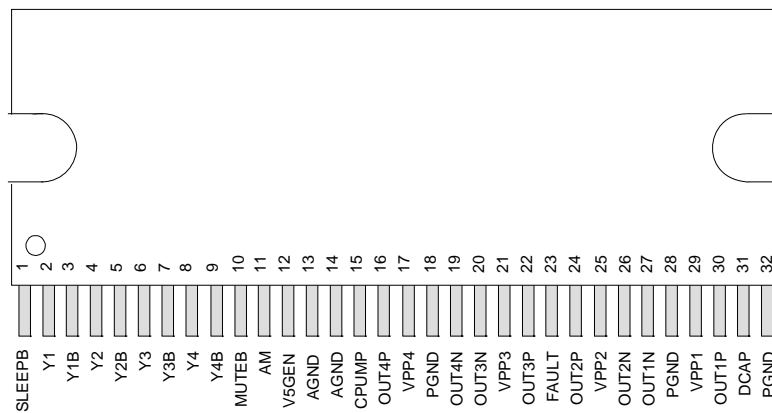
### Protection Circuits (Note 8)

T<sub>A</sub> = 25 °C. Unless otherwise noted, the supply voltage is V<sub>PP</sub>=14.4V.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNITS
OV <sub>ON</sub>	Over-voltage Threshold	Over-voltage turn on (amp muted)	27.0	30	32.5	V
OV <sub>OFF</sub>	Over-voltage Reset	Over-voltage turn off (mute off)	26.0	28.0		V
UV <sub>OFF</sub>	Under-voltage Reset	Under-voltage turn off (mute off)		9.5	10.0	V
UV <sub>ON</sub>	Under-voltage Threshold	Under-voltage turn on (amp muted)	7.8	8.1	8.6	V
OT <sub>ON</sub>	Over-Temperature Threshold	Over-temperature turn on (amp muted)	150	160	170	°C
OT <sub>OFF</sub>	Over-Temperature Reset	Over-temperature turn off (mute off)	120	130	140	°C
I <sub>OC</sub>	Over-Current Detect	1kHz single-shot ramp, V <sub>PP</sub> = 25V	10.5	12.5		A
V <sub>P</sub> MAX	Load Dump Voltage Withstand	Test conditions, t <sub>r</sub> > 2.5ms, t <sub>pulse</sub> < 50ms	60			V

### TPS4100A Pinout

32-pin SSIP Package  
(Top view)

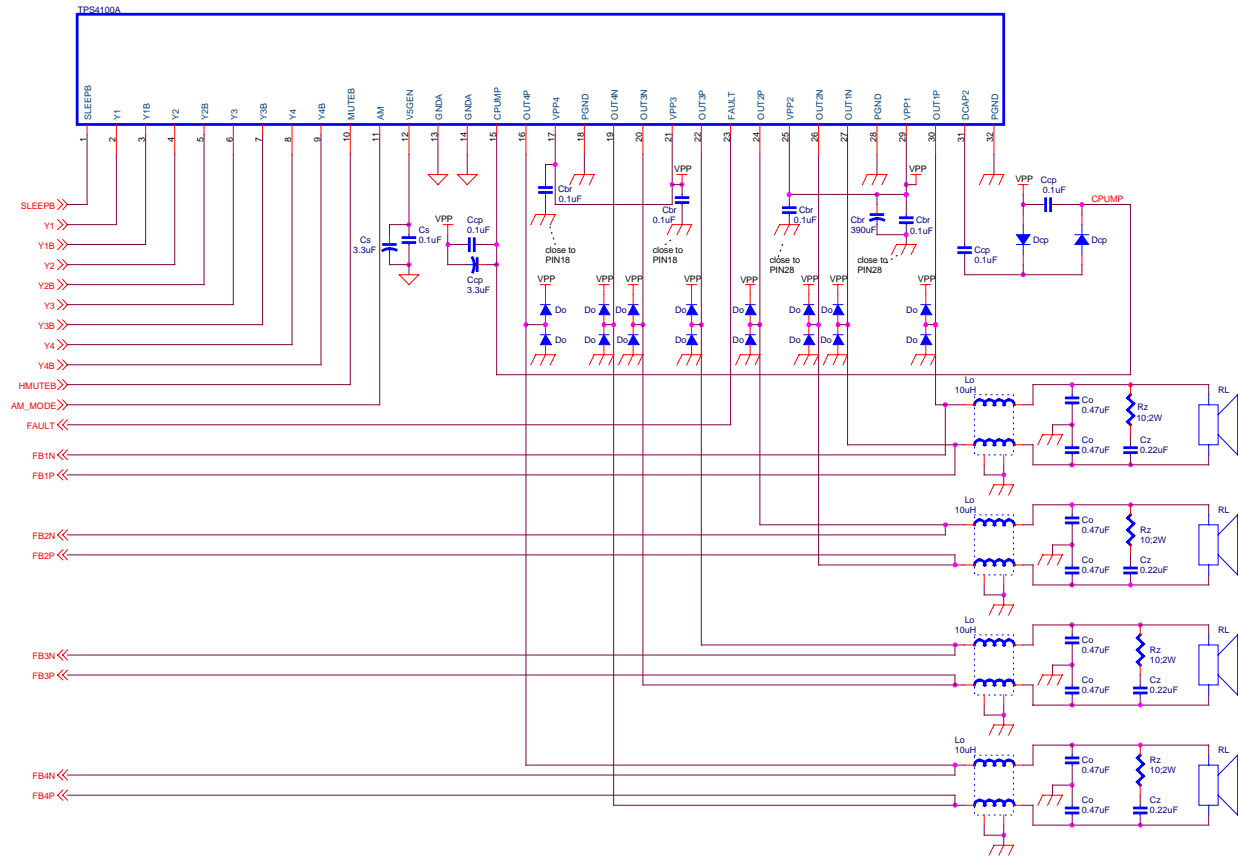


Note: The heat slug of the TPS4100A is connected to PGND.

## TPS4100A Pinout

PIN	NAME/FUNCTION	TYPE	DESCRIPTION
1	SLEEPB	INPUT (L)	Logic input, ACTIVE LOW. Setting SLEEP to low puts the TPS4100A in sleep mode. Input range is 0 to 5V with 3.3V compliant inputs.
2, 4, 6, 8	Y1, Y2, Y3, Y4	INPUT (L)	Non-inverted switching modulator inputs.
3, 5, 7, 9	Y1B, Y2B, Y3B, Y4B	INPUT (L)	Inverted switching modulator inputs.
10	MUTEB	INPUT (L)	Logic Input, ACTIVE LOW. Setting MUTE to low puts the device in mute mode. Typically driven by external power supply or microcontroller. Input range is 0 to 5V with 3.3V compliant inputs.
11	AM	INPUT (L)	Logic input, ACTIVE HIGH. Enables Analog Mode operation. Typically driven by Tripath controller. Input range is 0 to 5V with 3.3V compliant inputs.
12	5VGEN	OUTPUT	On chip 5V regulator bypass capacitor connection
12	HMUTEB	OUTPUT (L)	Logic output, ACTIVE LOW. HMUTEB low indicates TPS4100A is in mute mode
13, 14	AGND	GND	Analog ground
15	CPUMP	OUTPUT	Charge pump output capacitor
16	OUT4P	OUTPUT	Positive Output Channel 4
17	VPP4	POWER	Positive Supply Voltage Channel 4
18	PGND	GND	Power Ground for Outputs 3 and 4
19	OUT4N	OUTPUT	Negative Output Channel 4
20	OUT3N	OUTPUT	Negative Output Channel 3
21	VPP3	POWER	Positive Supply Voltage Channel 3
22	OUT3P	OUTPUT	Positive Output Channel 3
23	FAULT	OUTPUT (L)	Open Drain Logic Output, ACTIVE HIGH. FAULT high indicates fault condition.
24	OUT2P	OUTPUT	Positive Output Channel 2
25	VPP2	POWER	Positive Supply Voltage Channel 2
26	OUT2N	OUTPUT	Negative Output Channel 2
27	OUT1N	OUTPUT	Negative Output Channel 1
28	PGND	GND	Power Ground for Outputs 1 and 2
29	VPP1	POWER	Positive Supply Voltage Channel 1
30	OUT1P	OUTPUT	Positive Output Channel 1
31	DCAP	OUTPUT	Oscillator output for driving external charge pump circuit
32	PGND	GND	Power Ground

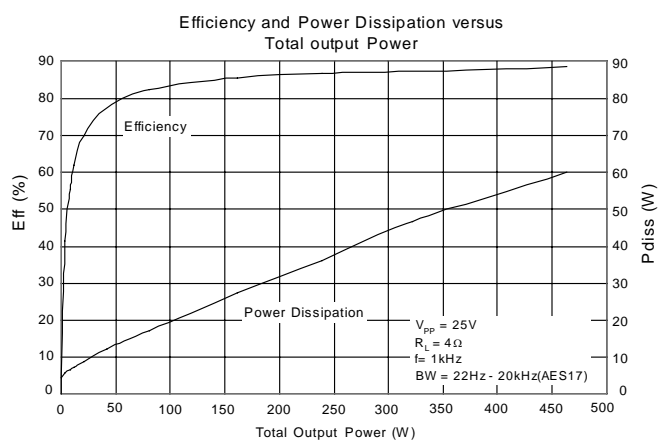
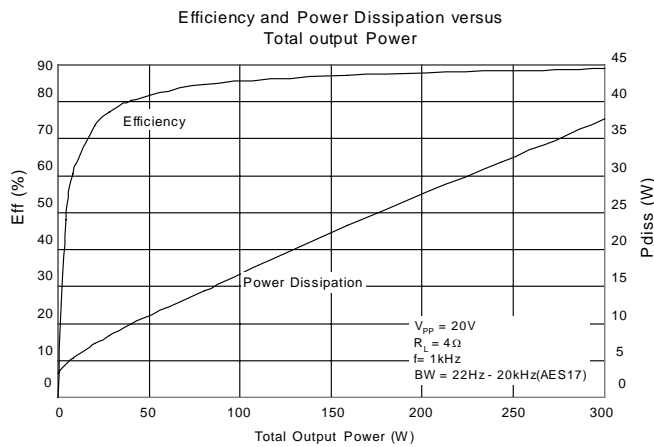
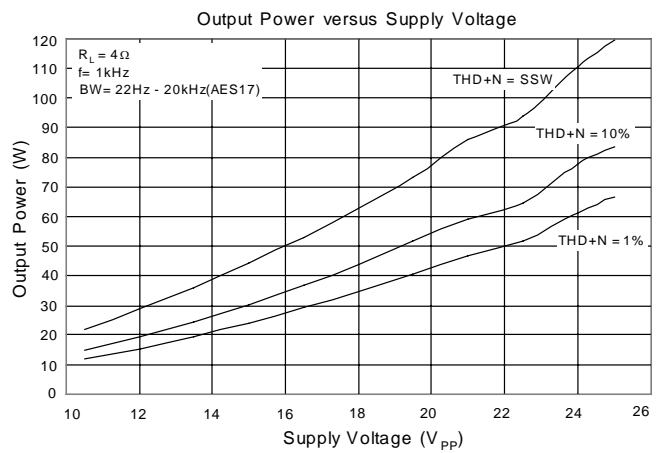
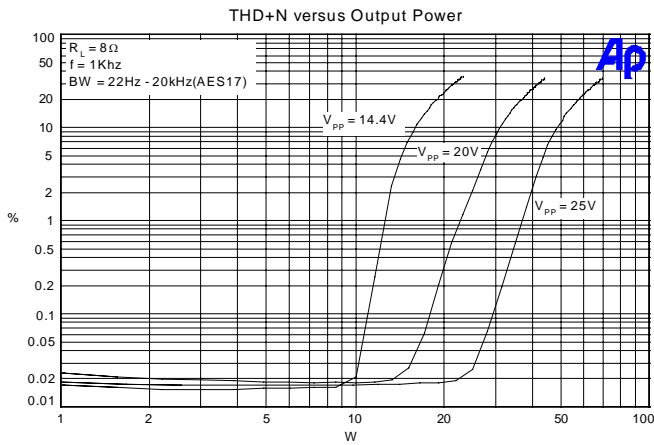
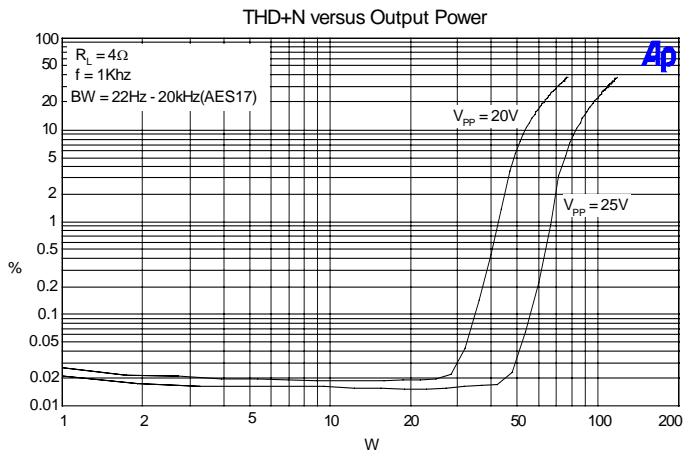
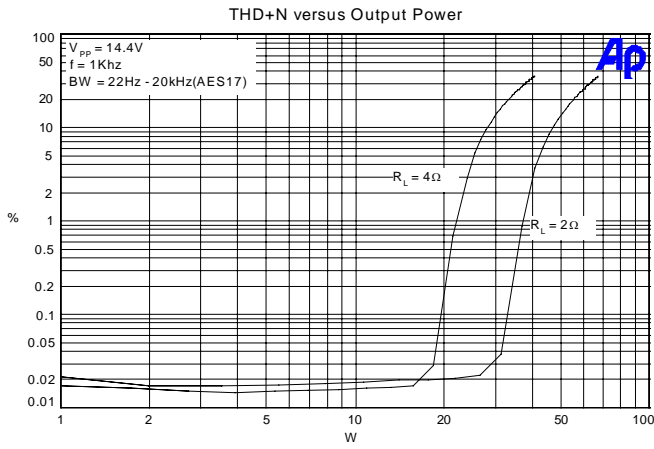
# TPS4100A Connection Diagram



**External Components Description** (Refer to the Connection Diagram)

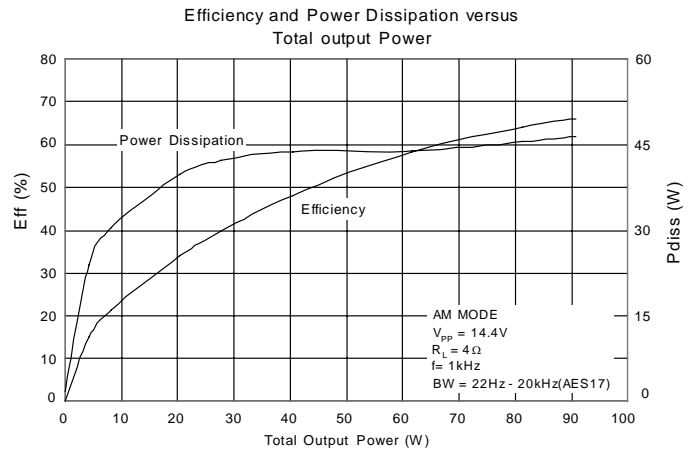
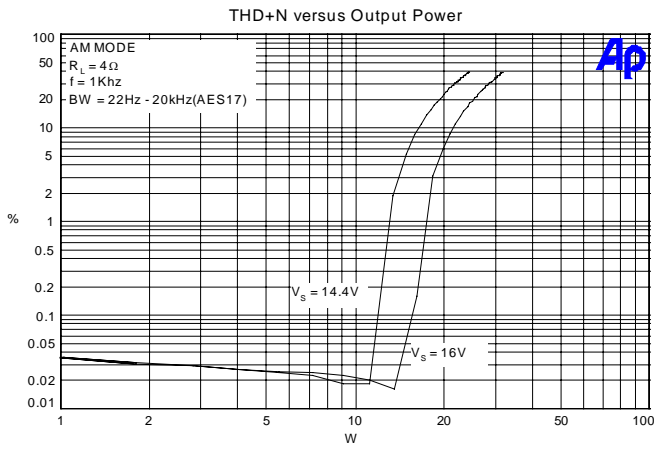
Components	Description
C <sub>S</sub>	Supply decoupling for the power supply pins. For optimum performance, these components should be located close to the TPS4100A and returned to their respective “ground” as shown in the Application/Test Circuit.
C <sub>BR</sub>	Supply decoupling for the high current full-bridge supply pins. These components must be located as close to the power supply pins as possible to minimize output ringing which causes power supply overshoot. By reducing overshoot, these capacitors maximize the TPS4100A reliability. These capacitors should have good high frequency performance including low ESR and low ESL.
C <sub>CP</sub>	Supply decoupling for the charge pump (high side gate drive supply) circuitry. These components must be located as close to the TPS4100A as possible.
D <sub>O</sub>	Output diode, which is used to minimize output overshoots/undershoots on the output nodes. These devices clamp the output to the low impedance node formed by the close connection of C <sub>BR</sub> . Note the connection shown in the Application/Test Circuit. The “high side” diode protects the bottom side device from excessive BV <sub>DSS</sub> due to overshoots on the output node. The “bottom side” diode protects the top side device from excessive BV <sub>DSS</sub> due to undershoots on the output node. This device must be an ultra fast rectifier capable of sustaining the entire supply range and high peak currents.
C <sub>Z</sub>	Zobel capacitor, which in conjunction with R <sub>Z</sub> , terminates the output filter at high frequencies. Use a high quality film capacitor capable of sustaining the ripple current caused by the switching outputs.
R <sub>Z</sub>	Zobel resistor, which in conjunction with C <sub>Z</sub> , terminates the output filter at high frequencies. The combination of R <sub>Z</sub> and C <sub>Z</sub> minimizes peaking of the output filter under both no load conditions or with real world loads, including loudspeakers, which usually exhibit a rising impedance with increasing frequency. Depending on the program material, the power rating of R <sub>Z</sub> may need to be adjusted. If the system requires full power operation at 20kHz then the power rating for R <sub>Z</sub> will likely need to be increased.
L <sub>O</sub>	Output inductor, which in conjunction with C <sub>O</sub> and C <sub>DO</sub> , demodulates (filters) the switching waveform into an audio signal. Forms a second order filter with a cutoff frequency of $f_c = 1 / (2\pi \sqrt{L_o C_{TOT}})$ and a quality factor of $Q = R_L C_{TOT} / 2\sqrt{L_o C_{TOT}}$ where C <sub>TOT</sub> = C <sub>O</sub> .
C <sub>O</sub>	Output capacitor, which, in conjunction with L <sub>O</sub> , demodulates (filters) the switching waveform into an audio signal. Use a high quality film capacitor capable of sustaining the ripple current caused by the switching outputs.
R <sub>FAULT</sub>	Pull-up resistor for the open drain FAULT pin output. Recommended resistor value is 51kΩ.
D <sub>CP</sub>	Charge pump diodes. Used to generate floating supply for driving high side circuitry. Small signal diodes such as 1N4148 are recommended for these components.

## Typical Performance





## Typical Performance AM Mode



## Application Information

### GENERAL DESCRIPTION

The TPS4100A is a 4-channel BTL (Bridge Tied Load) audio amplifier power stage that operates on a single supply voltage ranging from 10-26V. The device is targeted specifically to meet the demands of OEM and aftermarket automobile in-dash head units. With a single supply voltage of 25V, the device delivers four 100 Watt (saturated square wave) channels into 4 ohm. Since the TPS4100A is a switching amplifier, the average dissipation at low to medium output power is far superior to best in class AB amplifiers specifically designed for in-dash head units.

### POWER SUPPLY REQUIREMENTS

The device is configured to operate from a single supply voltage of 10-26V. This allows the device to operate from an automobile battery under various conditions including: battery voltage with the engine off, alternator voltage with engine running and boosted voltage operation up to 26V using a DC-DC converter or voltage booster. The sleep pin must be driven from a microcontroller or external 3.3V or 5.0V power supply.

### AM MODE

The TPS4100A is typically configured as a high power, high efficiency, four channel switching amplifier. The TPS4100A also has an additional amplifier mode named “AM Mode.” By pulling the AM pin to a logic high level, the TPS4100A is configured as a Class B amplifier as opposed to the normal, Class-T amplifier.

AM mode significantly reduces EMI generation since the output amplifiers are now operated in linear mode. Operating in Class B mode also reduces the TPS4100A efficiency especially at low to medium output powers. Due to this increased power dissipation, it is recommended that the AM mode is used for applications such as AM radio playback where the average output level is minimal and a switching amplifier would most effect radio reception. The operating supply range is also limited to 10V-16V due to the increased power dissipation.

A DC to DC converter that is used to step up the battery voltage for Class T operation (for instance 14.4V → 20V or above) should be disabled before entering AM Mode. This will ensure that the supply voltage in AM Mode is limited to the battery voltage thereby minimizing the device power dissipation.

To avoid possible damage to the output stage, appropriate sequencing must be adhered to when activating or disabling AM Mode. The TPS4100A **MUST** be muted (MUTE<sub>B</sub> = high) during logic changes of the AM pin.

To change from Class-T mode to AM Mode, the following procedure must be followed.

With AM = Low and MUTE<sub>B</sub> = High → MUTE<sub>B</sub> = Low → AM = High → MUTE<sub>B</sub> = High

To change from AM Mode to Class-T mode, the following procedure must be followed.

With AM = High, MUTE<sub>B</sub> = High → MUTE<sub>B</sub> = Low → AM = Low → MUTE<sub>B</sub> = High

Thus, MUTE<sub>B</sub> is low during any transition of the AM pin.

### CIRCUIT BOARD LAYOUT

When used in conjunction with a Class T controller the TPS4100A is a power (high current) amplifier that operates at relatively high switching frequencies. Therefore, amplifier outputs switch between the supply

voltage and ground at high speeds while driving high currents. This high-frequency digital signal is passed through an LC low-pass filter to recover the amplified audio signal. Since the amplifier must drive the inductive LC output filter and speaker loads, the amplifier outputs can be pulled above the supply voltage and below ground by the energy in the output inductance. To avoid subjecting the TPS4100A to potentially damaging voltage stress, it is critical to have a good printed circuit board layout. It is recommended that Tripath's layout and application circuit be used for all applications and only be deviated from after careful analysis of the effects of any changes.

The following components are important to place near either their associated TPS4100A pins. The recommendations are ranked in order of layout importance, either for proper device operation or performance considerations.

- The capacitors,  $C_{BR}$ , provide high frequency bypassing of the amplifier power supplies and will serve to reduce spikes and modulation of the power supply rails. Please note that bypassing requires a combination of capacitors for adequate stabilization.
- The output diodes,  $D_O$ , are used to minimize overshoots/undershoots on the output node. Improper routing of these diodes will render them useless due to PCB trace inductance. Thus, these components must be located very close to the output pins with the "other side of the diode" routed directly to the appropriate VPP or PGND pin.

The capacitors,  $C_S$ , provide high frequency bypassing of the amplifier power supplies. Please note that bypassing requires a combination of capacitors for adequate stabilization.

## **PROTECTION CIRCUITS**

The TPS4100A is guarded against over-current, over/under voltage, and over-temperature conditions. If the device goes into one of the various protection states, the FAULT pin goes to a logic HIGH state indicating a fault condition. When this occurs, all amplifier outputs are TRI-STATED and will float to VDD.

### **OVER-CURRENT PROTECTION**

An over-current fault occurs if more than approximately 12.5 amps (typical) of current flows from any of the amplifier output pins. This can occur if the speaker wires are shorted together, if one side of the speaker is shorted to ground, or if an output is connected to VPP.

### **OVER AND UNDER VOLTAGE PROTECTION**

An over-voltage fault occurs if the supply voltage is increased above 30.0 volts (typical), 27.0 volts (minimum). This fault puts the amplifier into mute and resets automatically once the supply voltage is reduced below the hysteresis band. The TPS4100A also has built-in load dump protection. This circuit puts the amplifier into sleep mode if the supply voltage is increased above 30V. The TPS4100A is able to survive power supply spikes to 60V if the duration is less than 50mS.

The TPS4100A is also equipped with under voltage protection. This circuit is activated if the supply voltage goes below 8.1 volts (typical) and causes the output to mute. Increasing the supply voltage above the hysteresis band (typically 9.5V) will bring the amplifier out of mute mode.

### **OVER-TEMPERATURE PROTECTION**

An over-temperature FAULT occurs if the junction temperature of the part exceeds 160°C (typical). The thermal hysteresis is approximately 30°C, therefore the fault will automatically clear when the junction temperature drops below 130°C.

## **SLEEP PIN (ACTIVE LOW)**

The SLEEPB ( $\overline{\text{SLEEP}}$ ) pin is a logic input that when pulled low puts the TPS4100A into a low quiescent current mode. This pin must be pulled up to an external 3.3V or 5V supply to activate (disable sleep mode) the TPS4100A. The sleep pin cannot be pulled up to VPP due to internal circuitry limitations. The amplifier takes approximately 500mS to come out of sleep. This period of time allows the input capacitor to charge fully assuming a value of 0.47 $\mu$ F. If the input capacitor size is increased, then additional time will be required to allow for the input capacitor to fully charge. To ensure that turn on is pop-free, the input capacitor must be fully charged before MUTE $\overline{\text{B}}$  is pulled high.

## **MUTE $\overline{\text{B}}$ PIN**

The MUTE $\overline{\text{B}}$  pin is a logic input that mutes the TPS4100A. Pulling this pin low activates the mute circuitry. Pulling the pin high enables output switching and amplification. Please note that the input stage is still biased at approximately 2.5V, even when MUTE $\overline{\text{B}}$  pin is low. This keeps the BIASCAP, C $\overline{\text{B}}$  and input coupling capacitors, C $\text{I}$ , completely charged. This allows for a clean transition from mute to on, and vice-versa, which eliminates turn-on/off pops. Please note that DC calibration is done every time MUTE $\overline{\text{B}}$  transitions from low to high. The DC calibration takes approximately 6mS.

## **FAULT PIN**

The FAULT pin is a logic output that indicates various fault conditions within the device. These conditions include: over-voltage, under-voltage, over-current at any output, low charge pump voltage, low 5V regulator voltage, and over-temperature (junction temperature greater than approximately 160°C).

The FAULT pin is an open drain output. The recommended pull-up to an external 3.3V or 5V supply is 51k $\Omega$ . Alternatively, this pin can be pulled up to VPP through a 51k $\Omega$  resistor. A logic high on this pin indicates a fault condition. This pin has a 1mA maximum sink current capability.

## **OUTPUT FILTER DESIGN**

One advantage of Tripath amplifiers over PWM solutions is the ability to use higher-cutoff-frequency filters. This means any load-dependent peaking/droop in the 20kHz audio band potentially caused by the filter can be made negligible. This is especially important for applications where the user may select a 4 $\Omega$  or 8 $\Omega$  speaker. Furthermore, speakers are not purely resistive loads and the impedance they present changes over frequency and from speaker model to speaker model.

The core material of the output filter inductor has an effect on the distortion levels produced by a TPS4100A amplifier. Tripath recommends low- $\mu$  type-2 iron powder cores because of their low loss and high linearity or high current capability bobbin types that will not saturate at peak currents below 9A.

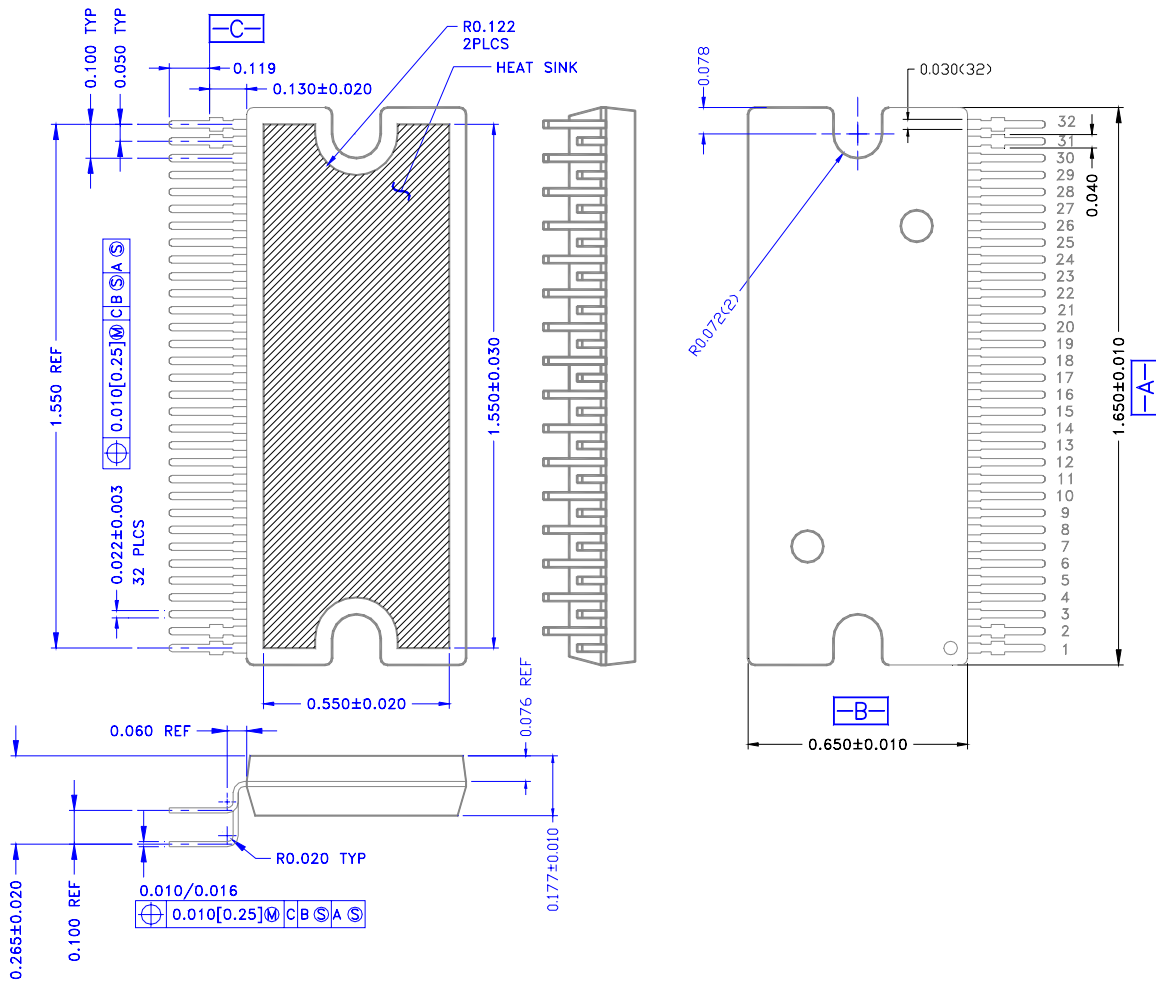
Recently, there have been a number of dual inductors designed specifically for bridged output switching amplifiers. These dual inductors are two inductors shielded by a common ferrite shield. They may be manufactured as common mode chokes with the windings wound in the same direction or as differential mode chokes with the windings wound in opposite directions. Since the ferrite reduces the energy storage capability of the inductor, it is important to ensure that the shielded dual inductor does not saturate at the maximum currents attainable by the TPS4100A. Dual inductors wound as common mode inductors may aid in reducing common mode noise to the load. They also may result in lower than initial inductances due to electric field cancellation effects.

Tripath also recommends that an RC damper be used after the LC low-pass filter. No-load operation of a TPS4100A amplifier can create significant peaking in the LC filter, which produces strong resonant

currents that can overheat the integrated MOSFETs and/or other components. The RC dampens the peaking and prevents problems.

It is highly recommended that the design process for a TPS4100A amplifier include an analysis of the interaction of intended speaker(s) with the LC filter and RC damper to ensure the desired frequency response is attained. Component values for the LC filter and RC damper may need to be altered from the Tripath suggestions to achieve the required response.

## Package Information



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