TOSHIBA CMOS Integrated Circuit Silicon Monolithic

## TB6575FNG

PWM Sensorless Controller for 3-Phase Full-Wave BLDC Motors

The TB6575F NG provides sensorless commutation and PWM current control for 3 -phase full-wave BLDC motors. It controls rotation speed by changing a PWM duty cycle by analog voltage.

## Features

- 3-phase full-wave sensorless drive
- PWM chopper drive
- PWM duty cycle control by analog input
- 20-mA current sink capability on PWM output pins
- Overcurrent protection
- Forward/reverse rotation


Weight: 0.14 g (typ.)

- Lead angle control $\left(7.5^{\circ}\right.$ and $\left.15^{\circ}\right)$
- Overlap commutation
- Rotation speed sensing signal
- DC excitation mode to improve startup characteristic
- DC excitation time and forced commutation time for startup operation can be changed.
- F orced commutation frequency can be selected. (fXT/(6×216), fXT/(6 $\left.\times 2^{17}\right), \mathrm{fXT} /\left(6 \times 2^{18}\right)$ )
- Output polarity switching (P-channel +N -channel, N -channel +N -channel)

The following conditions apply to solderability: *Solderability

1. Use of Sn-37Pb solder bath
*solder bath temperature $=230^{\circ} \mathrm{C}$
*dipping time $=5$ seconds
*number of times = once
*use of R-type flux
2. Use of Sn-3.0Ag-0.5Cu solder bath
*solder bath temperature $=245^{\circ} \mathrm{C}$
*dipping time $=5$ seconds
*number of times = once
*use of R-type flux

## Block Diagram



## Pin Assignment



## Pin Description

| Pin No. | Symbol | I/O | Description |
| :---: | :---: | :---: | :---: |
| 1 | GND | - | Ground pin |
| 2 | SC | I | Connection pin for a capacitor to set a startup commutation time and duty cycle ramp-up time |
| 3 | OS | 1 | Select the polarity of transistors. <br> High or open: High-side transistor $=$ P-channel (active low) Low-side transistor $=\mathrm{N}$-channel (active low) <br> Low $\quad:$ High-side transistor $=\mathrm{N}$-channel (active low) <br> Low-side transistor = N-channel (active low) <br> The pin has a pull-up resistor. |
| 4 | $F_{\text {MAX }}$ | 1 | Set an upper limit of the maximum commutation frequency. <br> <Fst=Low> <br> $\mathrm{F}_{\text {MAX }}=$ High or Open, Maximum commutation frequency $\mathrm{f}_{\mathrm{MX}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{11}\right)$ <br> $\mathrm{F}_{\mathrm{MAX}}=$ Low, Maximum commutation frequency $\mathrm{f}_{\mathrm{MX}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{12}\right)$ <br> <Fst=High or Middle> <br> $F_{M A X}=$ High or Open , Maximum commutation frequency $f_{M X}=f_{X T} /\left(6 \times 2^{8}\right)$ <br> $\mathrm{F}_{\text {MAX }}=$ Low, Maximum commutation frequency $\mathrm{f}_{\mathrm{MX}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{9}\right)$ <br> The pin has a pull-up resistor. |
| 5 | $\mathrm{V}_{\text {SP }}$ | I | Duty cycle control input <br> $0 \leq \mathrm{V}_{\mathrm{SP}} \leq \mathrm{V}_{\mathrm{AD}}(\mathrm{L})$ : Output off <br> $\mathrm{V}_{\text {AD }}(\mathrm{L}) \leq \mathrm{V}_{\mathrm{SP}} \leq \mathrm{V}_{\text {AD }}(\mathrm{H})$ : Set the PWM duty cycle according to the analog input. <br> $\mathrm{V}_{\text {AD }}(\mathrm{H}) \leq \mathrm{V}_{\mathrm{SP}} \leq \mathrm{V}_{\mathrm{DD}}:$ Duty cycle $=100 \%(31 / 32)$ <br> The pin has a pull-down resistor. |
| 6 | CW_CCW | I | Rotation direction input <br> High : Reverse rotation $(\mathrm{U} \rightarrow \mathrm{W} \rightarrow \mathrm{V})$ <br> Low or open : Forward rotation ( $\mathrm{U} \rightarrow \mathrm{V} \rightarrow \mathrm{W}$ ) <br> The pin has a pull-down resistor. |
| 7 | FG_OUT | 0 | Rotation speed sensing output <br> The pin is low at startup or upon a detection of a fault. This pin drives three pulses per rotation (3 ppr) based on the back-EMF (electromotive force) sensing. (In the case of 4 pole motor, 6 pulse output per rotation.) |
| 8 | START | 0 | DC excitation time setting pins |
| 9 | IP | 1 | After the IP pin reaches $\mathrm{V}_{\mathrm{DD}} / 2$, the TB6575FNG moves from DC excitation to forced commutation mode. |
| 10 | $\mathrm{X}_{\mathrm{T}}$ | - | Connection pins for a crystal oscillator |
| 11 | $\mathrm{X}_{\text {Tin }}$ | - | These pins have a feedback resistor. |
| 12 | LA | 1 | Lead angle control input <br> LA $=$ Low or open : Lead angle of $7.5^{\circ}$ <br> $\mathrm{LA}=$ high $\quad$ : Lead angle of $15^{\circ}$ <br> The pin has a pull-down resistor. |
| 13 | OUT_UP | O | PWM output signal for the high-side (positive-side) transistor driving motor phase $U$ The PWM polarity can be specified by pin 3. |
| 14 | OUT_UN | O | PWM output signal for the low-side (negative-side) transistor driving motor phase U This signal is active high. |
| 15 | OUT_VP | O | PWM output signal for the high-side (positive-side) transistor driving motor phase V The PWM polarity can be specified by pin 3. |
| 16 | OUT_VN | O | PWM output signal for the low-side (negative-side) transistor driving motor phase V This signal is active high. |
| 17 | OUT_WP | O | PWM output signal for the high-side (positive-side) transistor driving motor phase W The PWM polarity can be specified by pin 3. |
| 18 | OUT_WN | O | PWM output signal for the low-side (negative-side) transistor driving motor phase W This signal is active high. |
| 19 | Duty | O | PWM output monitor pin <br> This pin drives PWM output whose duty cycle corresponds to the $\mathrm{V}_{\mathrm{SP}}$ input. It also reflects the information at the OC pin. |
| 20 | SEL_LAP | 1 | Overlap commutation select pin Low: Overlap commutation High: $120^{\circ}$ commutation The pin has a pull-up resistor. |


| Pin No. | Symbol | I/O | Description |
| :---: | :---: | :---: | :---: |
| 21 | $V_{\text {DD }}$ | - | 5-V power supply pin |
| 22 | OC | 1 | Overcurrent detection input <br> The all PWM output signals are stopped when OC $\geq 0.5(\mathrm{~V})$. The pin has a pull-up resistor. |
| 23 | WAVE | 1 | Position sensing input <br> 3-phase voltage is applied to this pin. The pin has a pull-up resistor. |
| 24 | FST | 1 | Forced commutation frequency select pin <br> High or open: Forced commutation frequency $\mathrm{f}_{\mathrm{ST}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{16}\right)$ <br> Middle : Forced commutation frequency $\mathrm{f}_{\mathrm{ST}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{17}\right)$ <br> Low : Forced commutation frequency $\mathrm{fST}_{\mathrm{ST}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{18}\right)$ <br> The pin has a pull-up resistor. |

## Functional Description

## 1. Sensorless drive

On receiving an analog voltage command input, the rotor is aligned to a known position in DC excitation mode, and then the rotation is started in forced commutation mode by applying a PWM signal to the motor. As the rotor moves, back-EMF is acquired.

When a signal indicating the polarity of each of the phase voltages including back-EMF is applied to the position signal input pin, automatic switching occurs from the forced commutation PWM signal to the natural commutation PWM signal (which is generated based on the back-EMF sensing) to drive a BLDC motor in sensorless mode.

## 2. Startup operation

When the motor is stationary, there is no back-EMF and the motor position is unknown. For this reason, the rotor is aligned to a known position in DC excitation mode and then the rotation is started in forced commutation mode. An external capacitor sets the times that the TB6575F NG stays in DC excitation and forced commutation modes. Those times vary depending on the motor type and motor loading. Thus, they must be adjusted experimentally.


The rotor is aligned to a known position in DC excitation mode for period (a), during which the IP pin voltage decreases to half VDD level. The time constant for the period is determined by $\mathrm{C}_{2}$ and $\mathrm{R}_{1}$. After that, switching occurs to forced commutation mode represented by (b). The duty cycles for DC excitation and forced commutation modes are determined according to the SC pin voltage. When the number of turn of a motor is time more than forced commutation frequency, the motor switches to sensorless mode. The duty cycle for sensorless mode is determined by the VSP value.

## 3. Forced commutation frequency

The forced commutation frequency for startup operation is set as follows.
The optimal frequency varies depending on the motor type and motor loading. Thus, It must be adjusted experimentally.

FST $=$ High or Open : Forced commutation frequency $\mathrm{fST}=\mathrm{fXT}_{\mathrm{X}} /\left(6 \times 2^{16}\right)$
FST $=$ Middle $\quad:$ F orced commutation frequency $\mathrm{fST}=\mathrm{fXT} /\left(6 \times{ }^{17}\right)$
FST $=$ Low $\quad:$ Forfend commutation frequency $\mathrm{fST}_{\mathrm{f}}=\mathrm{f}_{\mathrm{XT}} /\left(6 \times 2^{18}\right)$

* fXT: Crystal oscillator frequency


## 4. PWM frequency

The PWM frequency is determined by an external oscillator.
PWM frequency (fPWM) $=\mathrm{fXT}_{\mathrm{X}} / 256$

* fxt: Crystal oscillator frequency

The PWM frequency must be sufficiently high, compared with the electrical frequency of the motor and within the switching performance of the transistors.

OS = High or Open
PWM signal driving high-side transistors


PWM signal driving low-side transistors


## 5. Speed control VSP pin

An analog voltage applied to the V SP pin is converted by the 6-bit AD converter to control the duty cycle of the PWM.
$0 \leq \mathrm{V}_{\text {DUTY }} \leq \mathrm{VAD}_{\text {AD }}$ (L)
$\rightarrow$ Duty cycle $=0 \%$
$\mathrm{V}_{\text {AD }}(\mathrm{L}) \leq \mathrm{V}_{\text {DUTY }} \leq \mathrm{V}_{\text {AD }}(\mathrm{H})$
$\rightarrow$ Figure at the right (1/64 to 63/64)
$V_{A D}(H) \leq V_{D U T Y} \leq V_{D D}$
$\rightarrow$ Duty cycle $=100 \%$ (63/64)


## 6. Fault protection

When a signal indicating the following faults is applied to the WAVE pin, the output transistors are disabled. After about one second, the motor is restarted. This operation is repeated as long as a fault is detected.

- The maximum commutation frequency is exceeded.
- The rotation speed falls below the forced commutation frequency.



## 7. Motor position detection error

A position detection is synchronized with the PWM signal generated in the IC. Thus, a position detection error relative to the PWM signal frequency may occur. Keep this in mind especially when the TB6575F NG is used for a high-speed motor.
A detection is performed on the falling edge of the PWM signal. An error is recognized when the pin voltage exceeds the reference voltage.

Detection error time $<1 / f_{p} \quad f_{p}: P W M$ frequency $=f_{X T} / 256 \quad f_{X T}$ : Crystal oscillator frequency


## 8. Lead angle control

The motor runs with a lead angle of $0^{\circ}$ in forced commutation mode at startup. After switching to natural commutation, the lead angle automatically changes to the value set by the LA pin.


## 9. Overlap commutation

When SEL_LAP = high, the TB6575FNG is configured to allow for $120^{\circ}$ commutation. When SEL_LAP = low, it is configured to allow for overlap commutation. In overlap commutation, there is an overlap period during which both the outgoing transistor and incoming transistor are conducting (as shown in the shaded areas). This period varies according to the lead angle.


Absolute Maximum Ratings ( $\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$ )

| Characteristics | Symbol | Rating | Unit |
| :--- | :---: | :---: | :---: |
| Power supply voltage | $\mathrm{V}_{\text {DD }}$ | 5.5 | V |
| Input voltage | $\mathrm{V}_{\text {in }}$ | $-0.3 \sim \mathrm{~V}_{\mathrm{DD}}+0.3$ | V |
| Turn-on signal output current | I OUT | 20 | mA |
| Power dissipation | $\mathrm{P}_{\mathrm{D}}$ | $780(\mathrm{Note})$ | mW |
| Operating temperature | $\mathrm{T}_{\text {opr }}$ | $-30 \sim 105$ | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ | $-55 \sim 150$ | ${ }^{\circ} \mathrm{C}$ |

Note: Without a PCB, stand-alone operation
Recommended Operating Conditions ( $\mathbf{T a}=-\mathbf{3 0}$ to $105^{\circ} \mathrm{C}$ )

| Characteristics | Symbol | Test Condition | Min | Typ. | Max | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply voltage | $\mathrm{V}_{\mathrm{DD}}$ | - | 4.5 | 5.0 | 5.5 | V |
| Input voltage | $\mathrm{V}_{\text {in }}$ | - | -0.3 | - | $V_{D D}$ <br> +0.3 | V |
| Oscillation frequency | $\mathrm{f}_{\mathrm{XT}}$ | - | 2.0 | 4.0 | 8.0 | MHz |

Electrical Characteristics ( $\mathrm{Ta}=\mathbf{2 5}{ }^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ )

| Characteristics | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Static power supply current | IDD | - | $\mathrm{V}_{\mathrm{SP}}=0 \mathrm{~V}, \mathrm{X}_{\text {Tin }}=\mathrm{H}$ | - | 0.7 | 1 | mA |
| Dynamic power supply current | IDD (opr) | - | $\mathrm{V}_{\mathrm{SP}}=2.5 \mathrm{~V}, \mathrm{X}_{\mathrm{Tin}}=4 \mathrm{MHz},$ Output open | - | 2 | 6 | mA |
| Input current | IIN-1 (H) | - | $\mathrm{V}_{\mathrm{IN}}=5 \mathrm{~V}$, OC, WAVE, SEL_LAP $\mathrm{F}_{\text {MAX }}, \mathrm{F}_{\text {ST }}$, OS | - | 0 | 1 | $\mu \mathrm{A}$ |
|  | IIN-1 (L) | - | VIN $=0$ V, OC, WAVE, SEL_LAP, $\mathrm{F}_{\text {MAX }}$, $\mathrm{FST}_{\text {, }}$ OS | -75 | -50 | - |  |
|  | IIN-2 (H) | - | $\mathrm{V}_{\text {IN }}=5 \mathrm{~V}, \mathrm{CW}$ _CCW, LA, $\mathrm{V}_{\text {SP }}$ | - | 50 | 75 |  |
|  | IIN-2 (L) | - | $\mathrm{V}_{\text {IN }}=0 \mathrm{~V}, \mathrm{CW}$ _CCW, LA, $\mathrm{V}_{\text {SP }}$ | -1 | 0 | - |  |
| Input voltage | $\mathrm{V}_{\text {IN-1 }}(\mathrm{H})$ | - | OC, SEL_LAP, CW_CCW WAVE, LA, FMAX, OS | 3.5 | - | 5 | V |
|  | $\mathrm{V}_{\text {IN-1 }}(\mathrm{L})$ | - | OC, SEL_LAP, CW_CCW WAVE, LA, F MAX , OS | GND | - | 1.5 |  |
|  | VIN-2 (H) | - | FST | 4 | - | 5 |  |
|  | $\mathrm{V}_{\text {IN-2 }}(\mathrm{M})$ | - | FST | 2 | - | 3 |  |
|  | $\mathrm{V}_{\text {IN-2 }}(\mathrm{L})$ | - | FST | GND | - | 1 |  |
| Input hysteresis voltage | $\mathrm{V}_{\mathrm{H}}$ | - | WAVE, I P | - | 0.45 | - | V |
| Output voltage | $\mathrm{V}_{\mathrm{O}-1}(\mathrm{H})$ | - | $\begin{aligned} & \mathrm{IOH}=-2 \mathrm{~mA} \\ & \mathrm{OUT} \text { OUP, OUT_VP, OUT_WP } \end{aligned}$ | 4.5 | - | $V_{\text {DD }}$ | V |
|  | $\mathrm{V}_{\mathrm{O}-1}(\mathrm{~L})$ | - | $\begin{aligned} & \text { IOL=20 mA } \\ & \text { OUT_UP, OUT_VP, OUT_WP } \end{aligned}$ | GND | - | 0.5 |  |
|  | $\mathrm{V}_{\mathrm{O}-2}(\mathrm{H})$ | - | $\begin{aligned} & \mathrm{IOH}=-20 \mathrm{~mA} \\ & \mathrm{OUT} \text { OUN, OUT_VN, OUT_WN } \end{aligned}$ | 4.5 | - | $V_{\text {DD }}$ |  |
|  | $\mathrm{V}_{\mathrm{O}-2}(\mathrm{~L})$ | - | $\begin{aligned} & \text { loL = } 2 \mathrm{~mA} \\ & \text { OUT_UN, OUT_VN, OUT_WN } \end{aligned}$ | GND | - | 0.5 |  |
|  | $\mathrm{V}_{\mathrm{O}-3}(\mathrm{H})$ | - | $\begin{aligned} & \mathrm{IOH}=-0.5 \mathrm{~mA} \\ & \mathrm{FG} \_\mathrm{OUT} \end{aligned}$ | 4.5 | - | $V_{\text {DD }}$ |  |
|  | $\mathrm{V}_{\mathrm{O}-3}(\mathrm{~L})$ | - | $\begin{aligned} & \mathrm{lOL}=0.5 \mathrm{~mA} \\ & \mathrm{FG} \_\mathrm{OUT} \end{aligned}$ | GND | - | 0.5 |  |
| Output leak current | $\mathrm{I}_{\mathrm{L}}(\mathrm{H})$ | - | $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{OUT}}=0 \mathrm{~V}$ <br> OUT_UP, OUT_VP, OUT_WP, OUT_UN, OUT_VN, OUT_WN, FG_OUT | - | 0 | 10 | $\mu \mathrm{A}$ |
|  | L ( L ) | - | $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{OUT}}=5.5 \mathrm{~V}$ OUT_UP, OUT_VP, OUT_WP OUT_UN, OUT_VN, OUT_WN, FG_ŌUT | - | 0 | 10 |  |
| PWM input voltage | $\mathrm{V}_{\text {AD }}(\mathrm{L})$ | - | $\mathrm{V}_{\mathrm{SP}}$ | 0.8 | 1.0 | 1.2 | V |
|  | $\mathrm{V}_{\text {AD }}(\mathrm{H})$ |  |  | 3.8 | 4.0 | 4.2 |  |
| CSC charge current | ISC | - | SC | 2.6 | 3.8 | 5.0 | $\mu \mathrm{A}$ |
| Fault retry time | TOFF | - | $\mathrm{V}_{\mathrm{SP}}=4 \mathrm{~V}, \mathrm{SC}$ pin $=0.47 \mu \mathrm{~F}$ | - | 940 | - | ms |
| Overcurrent detection voltage | Voc | - | OC | 0.46 | 0.5 | 0.54 | V |

## Input Equivalent Circuit

1. VSP pin

2. LA and CW_CCW pins

3. $X_{\text {Tin }}$ and $X_{\text {Tout }}$ pins

4. SEL_LAP, FMAX, FST, WAVE and OS pins


Hysteresis width
WAVE : 450 mV (typ.)
4. OUT_UP, OUT_UN, OUT_VP, OUT_VN, OUT_WP, OUT_WN and FG_OUT pins

6. OC pin

TOSHIBA


## Package Dimensions



Weight: 0.14 g (typ.)

## Notes on Contents

## 1. Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

## 2. Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

## 3. Timing Charts

Timing charts may be simplified for explanatory purposes.

## 4. Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass production design stage.
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## 5. Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

## IC Usage Considerations

## Notes on handling of ICs

[1] The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.
Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
[2] Do not insert devices in the wrong orientation or incorrectly.
Make sure that the positive and negative terminals of power supplies are connected properly. Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
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